## INTEGRATED CIRCUITS

## DATA SHEET

# **TDA1029**Signal-sources switch

Product specification
File under Integrated Circuits, IC01

January 1980





**TDA1029** 

The TDA1029 is a dual operational amplifier (connected as an impedance converter) each amplifier having 4 mutually switchable inputs which are protected by clamping diodes. The input currents are independent of switch position and the outputs are short-circuit protected.

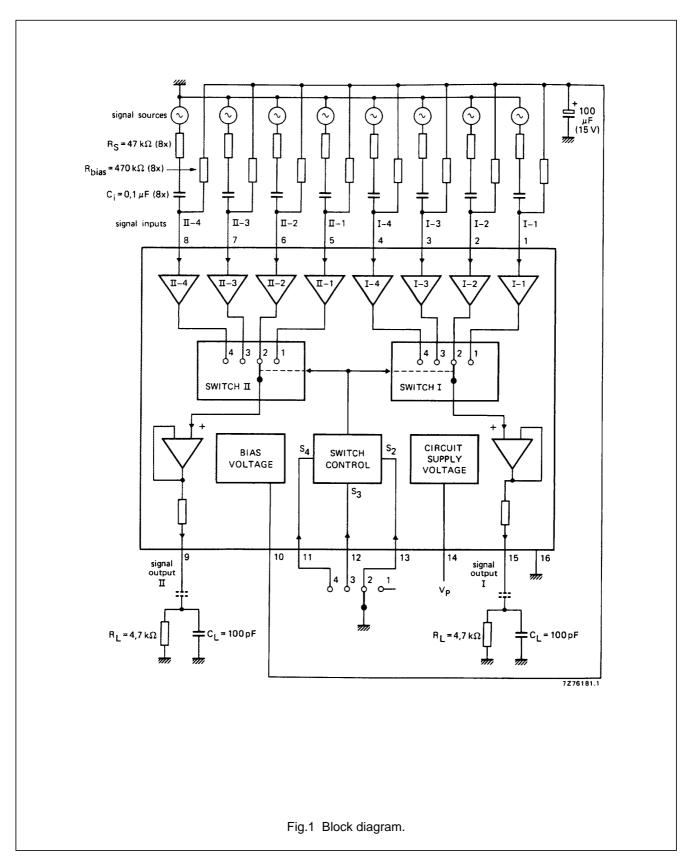
The device is intended as an electronic two-channel signal-source switch in a.f. amplifiers.

#### **QUICK REFERENCE DATA**

Supply voltage range (pin 14)	V <sub>P</sub>		6 to 23 V
Operating ambient temperature	$T_{amb}$	-3	30 to + 80 °C
Supply voltage (pin 14)	$V_{P}$	typ.	20 V
Current consumption	I <sub>14</sub>	typ.	3,5 mA
Maximum input signal handling (r.m.s. value)	$V_{i(rms)}$	typ.	6 V
Voltage gain	$G_v$	typ.	1
Total harmonic distortion	$d_{tot}$	typ.	0,01 %
Crosstalk	α	typ.	70 dB
Signal-to-noise ratio	S/N	typ.	120 dB

#### **PACKAGE OUTLINE**

16-lead DIL; plastic (SOT38); SOT38-1; 1996 July 18.



## Signal-sources switch

RATINGS	Out to the (IEO 404)		
Limiting values in accordance with the Absolute Maximum	,		
Supply voltage (pin 14)	$V_{P}$	max.	23 V
Input voltage (pins 1 to 8)	$V_{I}$	max.	$V_{P}$
	$-V_{I}$	max.	0,5 V
Switch control voltage (pins 11, 12 and 13)	$V_S$		0 to 23 V
Input current	$\pm I_I$	max.	20 mA
Switch control current	-I <sub>S</sub>	max.	50 mA
Total power dissipation	$P_{tot}$	max.	800 mW
Storage temperature	$T_{stg}$		−55 to + 150 °C
Operating ambient temperature	$T_{amb}$		−30 to + 80 °C
CHARACTERISTICS			
$V_P = 20 \text{ V}$ ; $T_{amb} = 25 ^{\circ}\text{C}$ ; unless otherwise specified			
Current consumption	1	typ.	3,5 mA
without load; $I_9 = I_{15} = 0$	I <sub>14</sub>		2 to 5 mA
Supply voltage range (pin 14)	$V_{P}$		6 to 23 V
Signal inputs			
Input offset voltage			
of switched-on inputs	V.	typ.	2 mV
$R_S \le 1 \text{ k}\Omega$	$V_{io}$	<	10 mV
Input offset current	ı	typ.	20 nA
of switched-on inputs	l <sub>io</sub>	<	200 nA
Input offset current			
of a switched-on input with respect to a	1	typ.	20 nA
non-switched-on input of a channel	l <sub>io</sub>	<	200 nA
Input bias current	ı	typ.	250 nA
independent of switch position	l <sub>i</sub>	<	950 nA
Capacitance between adjacent inputs	С	typ.	0,5 pF
D.C. input voltage range	$V_{I}$		3 to 19 V
Supply voltage rejection ratio; $R_S \le 10 \text{ k}\Omega$	SVRR	typ.	100 μV/V
Equivalent input noise voltage		•	·
R <sub>S</sub> = 0; f = 20 Hz to 20 kHz (r.m.s. value)	$V_{n(rms)}$	typ.	3,5 μV
Equivalent input noise current	11(11110)		•
f = 20 Hz to 20 kHz (r.m.s. value)	I <sub>n(rms)</sub>	typ.	0,05 nA
Crosstalk between a switched-on input	11(11113)	71	-,
and a non-switched-on input;			
measured at the output at $R_S = 1 \text{ k}\Omega$ ; $f = 1 \text{ kHz}$	α	typ.	100 dB
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Signal amplifier			
Voltage gain of a switched-on input			
at $I_9 = I_{15} = 0$ ; $R_L = \infty$	$G_v$	typ.	1
Current gain of a switched-on amplifier	G <sub>i</sub>	typ.	10 <sup>5</sup>
Signal outputs			
Output resistance (pins 9 and 15)	$R_o$	typ.	400 $\Omega$
Output current capability at $V_P = 6$ to 23 V	$\pm I_9$ ; $\pm I_{15}$	typ.	5 mA
Frequency limit of the output voltage			
$V_{i(p-p)}$ = 1 V; $R_S$ = 1 k $\Omega$ ; $R_L$ = 10 M $\Omega$ ; $C_L$ = 10 pF	f	typ.	1,3 MHz
Slew rate (unity gain); $\Delta V_{9-16}/\Delta t$ ; $\Delta V_{15-16}/\Delta t$			
$R_L = 10 \text{ M}\Omega; C_L = 10 \text{ pF}$	S	typ.	2 V/μs
Bias voltage			
D.C. output voltage	V	typ.	11 V <sup>(1)</sup>
	V <sub>10-16</sub>		10,2 to 11,8 V
Output resistance	R <sub>10-16</sub>	typ.	8,2 kΩ

#### **Switch control**

switched-on	interconnected		control voltages	
inputs	pins	V <sub>11-16</sub>	V <sub>12-16</sub>	V <sub>13-16</sub>
I-1, II-1	1-15, 5-9	Н	Н	Н
I-2, II-2	2-15, 6-9	Н	Н	L
I-3, II-3	3-15, 7-9	Н	L	Н
I-4, II-4	4-15, 8-9	L	Н	Н
I-4, II-4	4-15, 8-9	L	L	Н
I-4, II-4	4-15, 8-9	L	Н	L
I-4, II-4	4-15, 8-9	L	L	L
I-3, II-3	3-15, 7-9	Н	L	L

In the case of offset control, an internal blocking circuit of the switch control ensures that not more than one input will be switched on at a time. In that case safe switching-through is obtained at  $V_{SL} \le 1,5 \text{ V}$ .

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Required voltage			
HIGH	$V_{SH}$	>	$3,3 V^{(2)}$
LOW	VSL	<	2,1 V
Input current			
HIGH (leakage current)	I <sub>SH</sub>	<	1 μΑ
LOW (control current)	-I <sub>SL</sub>	<	250 μΑ

#### **Notes**

- 1.  $V_{10-16}$  is typically  $0.5 \cdot V_{14-16} + 1.5 \cdot V_{BE}$ .
- 2. Or control inputs open ( $R_{11,12,13-16} > 33 \text{ M}\Omega$ ).

## **APPLICATION INFORMATION**

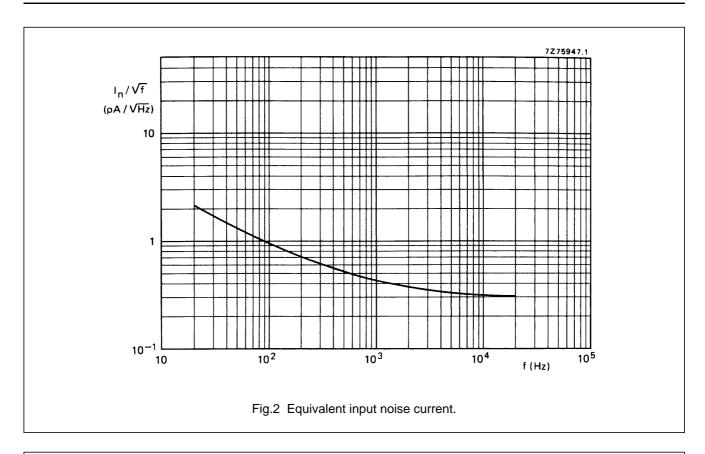
 $V_P$  = 20 V;  $T_{amb}$  = 25 °C; measured in Fig.1;  $R_S$  = 47 kΩ;  $C_i$  = 0,1 μF;  $R_{bias}$  = 470 kΩ;  $R_L$  = 47 kΩ;  $C_L$  = 100 pF (unless otherwise specified)

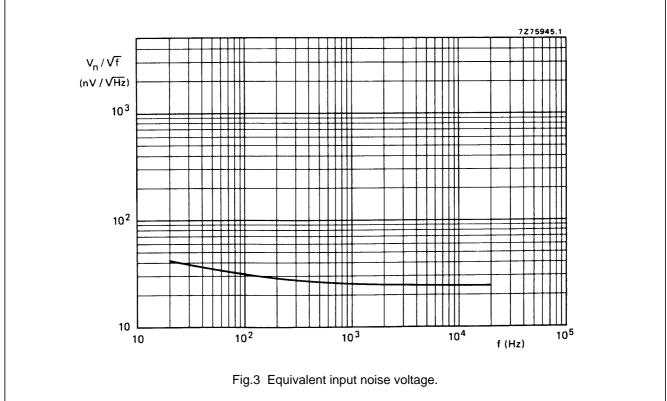
Voltage gain	$G_{v}$	typ.	-1,5	dB
Output voltage variation when switching	A\/ . A\/	typ.	10	mV
the inputs	$\Delta V_{9-16}; \Delta V_{15-16}$	<	100	mV
Total harmonic distortion				
over most of signal range (see Fig.4)	d <sub>tot</sub>	typ.	0,01	%
$V_i = 5 \text{ V}; f = 1 \text{ kHz}$	$d_{tot}$	typ.	0,02	%
$V_i = 5 \text{ V}$ ; $f = 20 \text{ Hz to } 20 \text{ kHz}$	d <sub>tot</sub>	typ.	0,03	%
Output signal handling				
d <sub>tot</sub> = 0,1%; f = 1 kHz (r.m.s. value)	V	>	5,0	V
$u_{\text{tot}} = 0,1\%, 1 = 1 \text{ KHZ (1.111.5. Value)}$	$V_{o(rms)}$	typ.	5,3	V
Noise output voltage (unweighted)				
f = 20 Hz to 20 kHz (r.m.s. value)	$V_{n(rms)}$	typ.	5	$\mu V$
Noise output voltage (weighted)				
f = 20 Hz to 20 kHz (in accordance with DIN 45405)	$V_n$	typ.	12	$\mu V$
Amplitude response				
$V_i$ = 5 V; f = 20 Hz to 20 kHz; $C_i$ = 0,22 $\mu F$	$\Delta V_{9-16;}  \Delta V_{15-16}$	<	0,1	dB <sup>(1)</sup>
Crosswalk between a switched-on input				
and a non-switched-on input;				
measured at the output at f = 1 kHz	α	typ.	75	dB <sup>(2)</sup>
Crosswalk between switched-on inputs				
and the outputs of the other channels	α	typ.	90	$dB^{(2)}$

#### **Notes**

- 1. The lower cut-off frequency depends on values of  $R_{\mbox{\scriptsize bias}}$  and  $C_{\mbox{\scriptsize i}}.$
- 2. Depends on external circuitry and R<sub>S</sub>. The value will be fixed mostly by capacitive crosstalk of the external components.

## Signal-sources switch





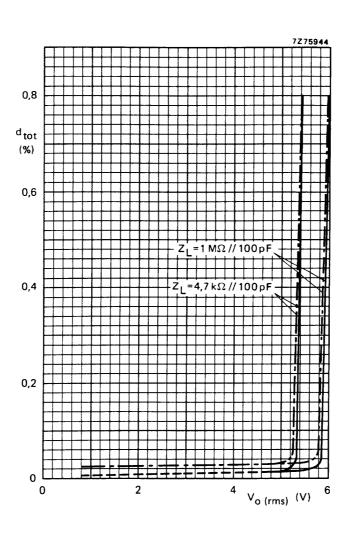
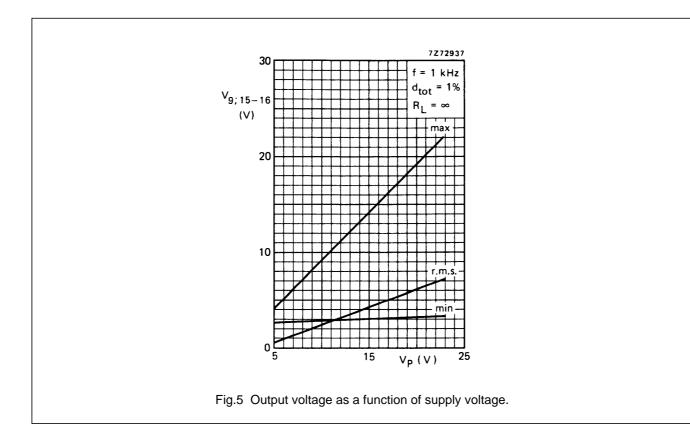
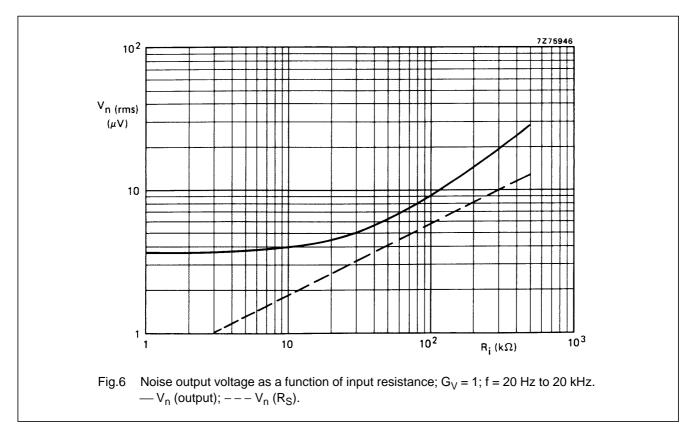


Fig.4 Total harmonic distortion as a function of r.m.s output voltage. — f = 1 kHz; — - — - f = 20 kHz.

## Signal-sources switch



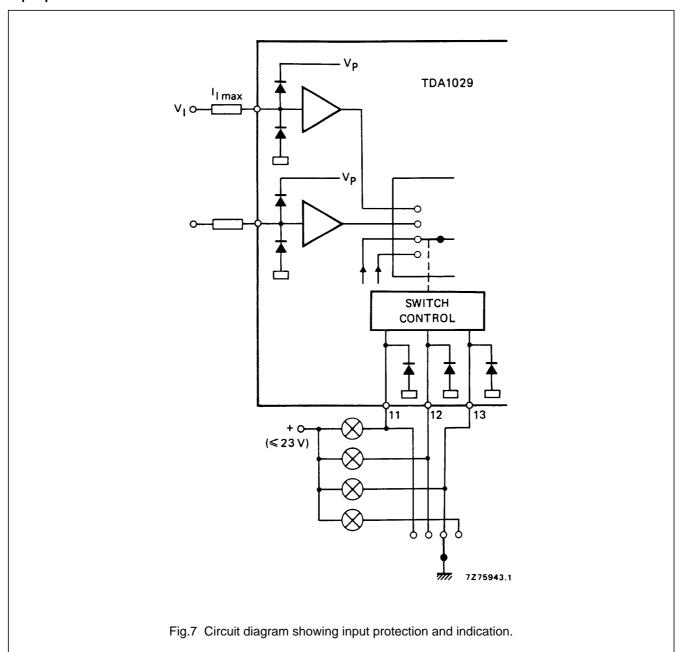


## Signal-sources switch

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#### **APPLICATION NOTES**

## Input protection circuit and indication

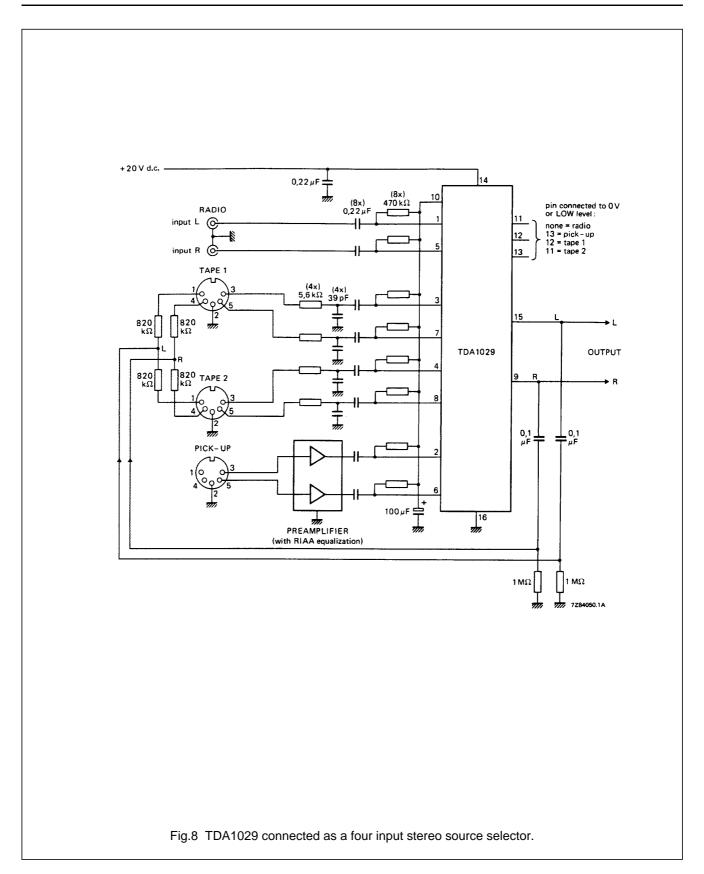


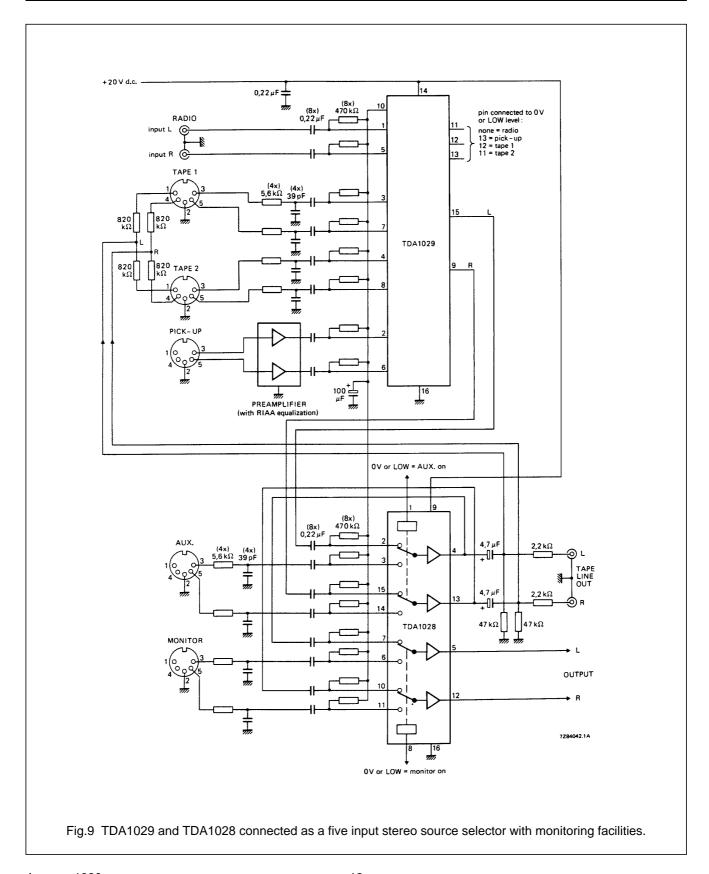
#### **Unused signal inputs**

Any unused inputs must be connected to a d.c. (bias) voltage, which is within the d.c. input voltage range; e.g. unused inputs can be connected directly to pin 10.

## Circuits with standby operation

The control inputs (pins 11, 12 and 13) are high-ohmic at  $V_{SH} \le 20 \text{ V}$  ( $I_{SH} \le 1 \text{ } \mu\text{A}$ ), as well as, when the supply voltage (pin 14) is switched off.





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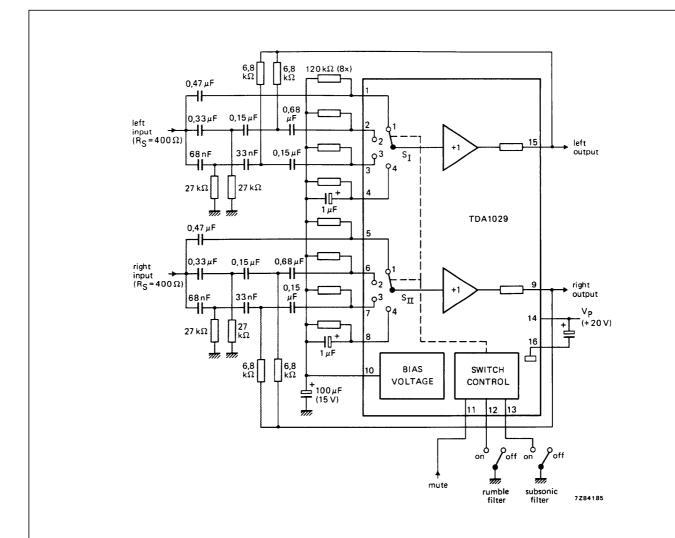


Fig.10 TDA1029 connected as a third-order active high-pass filter with Butterworth response and component values chosen according to the method proposed by Fjällbrant. It is a four-function circuit which can select mute, rumble filter, subsonic filter and linear response.

## **Switch control**

function	V <sub>11-16</sub>	V <sub>12-16</sub>	V <sub>13-16</sub>
linear	Н	Н	Н
subsonic filter 'on'	Н	Н	L
rumble filter 'on'	Н	L	X
mute 'on'	L	X	X

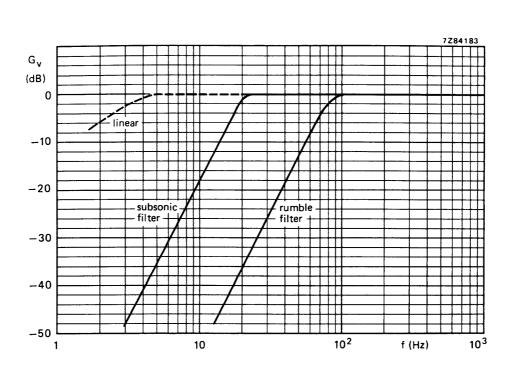


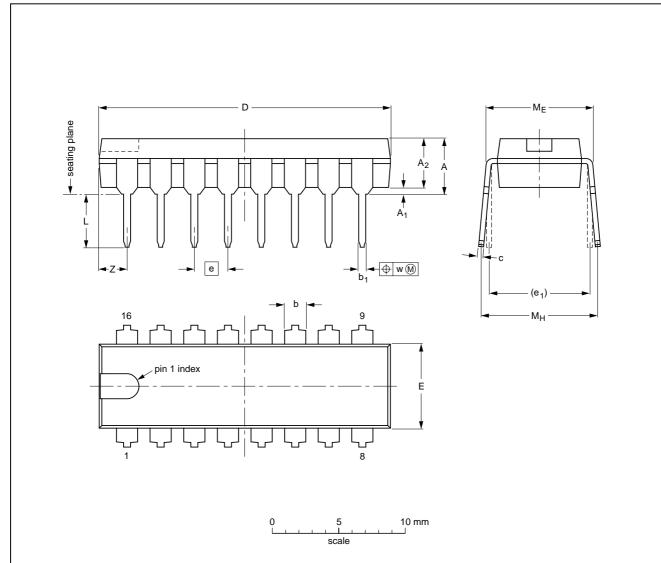
Fig.11 Frequency response curves for the circuit of Fig.10.

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## **PACKAGE OUTLINE**

DIP16: plastic dual in-line package; 16 leads (300 mil); long body

SOT38-1



## DIMENSIONS (inch dimensions are derived from the original mm dimensions)

UNIT	A max.	A <sub>1</sub> min.	A <sub>2</sub> max.	b	b <sub>1</sub>	С	D <sup>(1)</sup>	E <sup>(1)</sup>	е	e <sub>1</sub>	L	ME	Мн	w	Z <sup>(1)</sup> max.
mm	4.7	0.51	3.7	1.40 1.14	0.53 0.38	0.32 0.23	21.8 21.4	6.48 6.20	2.54	7.62	3.9 3.4	8.25 7.80	9.5 8.3	0.254	2.2
inches	0.19	0.020	0.15	0.055 0.045	0.021 0.015	0.013 0.009	0.86 0.84	0.26 0.24	0.10	0.30	0.15 0.13	0.32 0.31	0.37 0.33	0.01	0.087

#### Note

1. Plastic or metal protrusions of 0.25 mm maximum per side are not included.

OUTLINE		REFER	ENCES	EUROPEAN ISSUE DA		
VERSION	IEC	JEDEC	EIAJ	PROJECTION	ISSUE DATE	
SOT38-1	050G09	MO-001AE			<del>92-10-02</del> 95-01-19	

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#### **SOLDERING**

#### Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our "IC Package Databook" (order code 9398 652 90011).

## Soldering by dipping or by wave

The maximum permissible temperature of the solder is 260 °C; solder at this temperature must not be in contact with the joint for more than 5 seconds. The total contact time of successive solder waves must not exceed 5 seconds.

The device may be mounted up to the seating plane, but the temperature of the plastic body must not exceed the specified maximum storage temperature (T<sub>stg max</sub>). If the printed-circuit board has been pre-heated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

#### Repairing soldered joints

Apply a low voltage soldering iron (less than 24 V) to the lead(s) of the package, below the seating plane or not more than 2 mm above it. If the temperature of the soldering iron bit is less than 300 °C it may remain in contact for up to 10 seconds. If the bit temperature is between 300 and 400 °C, contact may be up to 5 seconds.

#### **DEFINITIONS**

Data sheet status	
Objective specification	This data sheet contains target or goal specifications for product development.
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.
Product specification	This data sheet contains final product specifications.
Limiting values	

#### Limiting values

Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.

#### **Application information**

Where application information is given, it is advisory and does not form part of the specification.

#### LIFE SUPPORT APPLICATIONS

These products are not designed for use in life support appliances, devices, or systems where malfunction of these products can reasonably be expected to result in personal injury. Philips customers using or selling these products for use in such applications do so at their own risk and agree to fully indemnify Philips for any damages resulting from such improper use or sale.